Introduction to Binary Convolutional Codes [1]

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Binary Convolutional Codes

- 1. A binary convolutional code is denoted by a three-tuple (n, k, m).
- 2. n output bits are generated whenever k input bits are received.
- 3. The current n outputs are linear combinations of the present k input bits and the previous $m \times k$ input bits.
- 4. m designates the number of previous k-bit input blocks that must be memorized in the encoder.
- 5. m is called the *memory order* of the convolutional code.

Encoders for the Convolutional Codes

- 1. A binary convolutional encoder is conveniently structured as a mechanism of shift registers and modulo-2 adders, where the output bits are modular-2 additions of selective shift register contents and present input bits.
- 2. n in the three-tuple notation is exactly the number of output sequences in the encoder.
- 3. k is the number of input sequences (and hence, the encoder consists of k shift registers).
- 4. *m* is the maximum length of the *k* shift registers (i.e., if the number of stages of the *j*th shift register is K_j , then $m = \max_{1 \le j \le k} K_j$).
- 5. $K = \sum_{j=1}^{k} K_j$ is the total memory in the encoder (K is sometimes called the *overall constraint lengths*).

6. The definition of constraint length of a convolutional code is defined in several ways. The most popular one is m + 1.

Encoder for the Binary (2, 1, 2) Convolutional Code \boldsymbol{u} is the information sequence and \boldsymbol{v} is the corresponding code sequence (codeword). $\mathbf{e} \mathbf{v}_1 = (1010011)$ u = (11101) $\mathbf{v} = (11\,01\,10\,01\,00\,10\,11)$ $- \boldsymbol{v}_2 = (1101001)$





1. The encoders of convolutional codes can be represented by *linear time-invariant* (LTI) systems.

2.

$$m{v}_j = m{u}_1 * m{g}_j^{(1)} + m{u}_2 * m{g}_j^{(2)} + \dots + m{u}_k * m{g}_j^{(k)} = \sum_{i=1}^k m{u}_i * m{g}_j^{(k)},$$

where * is the convolutional operation and $g_j^{(i)}$ is the impulse response of the *i*th input sequence with the response to the *j*th output.

- 3. $\boldsymbol{g}_{j}^{(i)}$ can be found by stimulating the encoder with the discrete impulse $(1, 0, 0, \ldots)$ at the *i*th input and by observing the *j*th output when all other inputs are fed the zero sequence $(0, 0, 0, \ldots)$.
- 4. The impulse responses are called *generator sequences* of the encoder.





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Generator Matrix in the Time Domain

- 1. The convolutional codes can be generated by a generator matrix multiplied by the information sequences.
- 2. Let u_1, u_2, \ldots, u_k are the information sequences and v_1, v_2, \ldots, v_n the output sequences.
- 3. Arrange the information sequences as

$$u = (u_{1,0}, u_{2,0}, \dots, u_{k,0}, u_{1,1}, u_{2,1}, \dots, u_{k,1}, \dots, u_{1,\ell}, u_{2,\ell}, \dots, u_{k,\ell}, \dots)$$

= $(w_0, w_1, \dots, w_\ell, \dots),$

and the output sequences as

$$v = (v_{1,0}, v_{2,0}, \dots, v_{n,0}, v_{1,1}, v_{2,1}, \dots, v_{n,1}, \dots, v_{1,\ell}, v_{2,\ell}, \dots, v_{n,\ell}, \dots)$$

= $(z_0, z_1, \dots, z_\ell, \dots)$

- 4. \boldsymbol{v} is called a codeword or code sequence.
- 5. The relation between \boldsymbol{v} and \boldsymbol{u} can characterized as

$$v = u \cdot G$$
,

where G is the generator matrix of the code.

6. The generator matrix is

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with the $k \times n$ submatrices

$$m{G}_\ell = \left(egin{array}{cccc} g_{1,\ell}^{(1)} & g_{2,\ell}^{(1)} & \cdots & g_{n,\ell}^{(1)} \ g_{1,\ell}^{(2)} & g_{2,\ell}^{(2)} & \cdots & g_{n,\ell}^{(2)} \ dots & dots$$

7. The element $g_{j,\ell}^{(i)}$, for $i \in [1, k]$ and $j \in [1, n]$, are the impulse response of the *i*th input with respect to *j*th output:

$$\boldsymbol{g}_{j}^{(i)} = (g_{j,0}^{(i)}, g_{j,1}^{(i)}, \dots, g_{j,\ell}^{(i)}, \dots, g_{j,m}^{(i)}).$$







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$$g_{2}^{(2)} = 4 \text{ (octal)}, g_{3}^{(1)} = 2 \text{ (octal)}, g_{3}^{(2)} = 3 \text{ (octal)}$$
$$v = u \cdot G$$
$$= (11,01) \cdot \begin{pmatrix} 100 & 001 & 000 \\ 010 & 001 & 001 \\ 100 & 001 & 000 \\ 010 & 001 & 001 \end{pmatrix}$$
$$= (110,010,000,001)$$

Generator Matrix in the Z Domain

1. According to the Z transform,

$$\boldsymbol{u}_{i} \quad \multimap \quad U_{i}(D) = \sum_{t=0}^{\infty} u_{i,t} D^{t}$$
$$\boldsymbol{v}_{j} \quad \multimap \quad V_{j}(D) = \sum_{t=0}^{\infty} v_{j,t} D^{t}$$
$$\boldsymbol{g}_{j}^{(i)} \quad \multimap \quad G_{i,j}(D) = \sum_{t=0}^{\infty} g_{j,t}^{(i)} D^{t}$$

2. The convolutional relation of the Z transform $Z\{u * g\} = U(D)G(D)$ is used to transform the convolution of input sequences and generator sequences to a multiplication in the Z domain.

3.
$$V_j(D) = \sum_{i=1}^k U_i(D) \cdot G_{i,j}(D).$$

4. We can write the above equations into a matrix multiplication:

$$\boldsymbol{V}(D) = \boldsymbol{U}(D) \cdot \boldsymbol{G}(D),$$

where

$$U(D) = (U_1(D), U_2(D), \dots, U_k(D))$$
$$V(D) = (V_1(D), V_2(D), \dots, V_n(D))$$
$$G(D) = \begin{pmatrix} G_{i,j} \end{pmatrix}$$



$$g_{1} = (1, 1, 1, 0, ...) = (1, 1, 1),$$

$$g_{2} = (1, 0, 1, 0, ...) = (1, 0, 1)$$

$$G_{1,1}(D) = 1 + D + D^{2}$$

$$G_{1,2}(D) = 1 + D^{2}$$

$$U_{1}(D) = 1 + D + D^{2} + D^{4}$$

$$V_{1}(D) = 1 + D^{2} + D^{5} + D^{6}$$

$$V_{2}(D) = 1 + D + D^{3} + D^{6}$$



$$\begin{aligned} \boldsymbol{g}_{1}^{(1)} &= (1,0,0), \ \boldsymbol{g}_{1}^{(2)} = (0,0,0), \ \boldsymbol{g}_{2}^{(1)} = (0,0,0), \\ \boldsymbol{g}_{2}^{(2)} &= (1,0,0), \ \boldsymbol{g}_{3}^{(1)} = (0,1,0), \ \boldsymbol{g}_{3}^{(2)} = (0,1,1), \\ G_{1,1}(D) &= 1, \ G_{1,2}(D) = 0, \ G_{1,3}(D) = D \\ G_{2,1}(D) &= 0, \ G_{2,2} = 1, \ G_{2,3}(D) = D + D^{2} \\ U_{1}(D) &= 1, \ U_{2}(D) = 1 + D \\ V_{1}(D) &= 1, \ V_{2}(D) = 1 + D, \ V_{3}(D) = D^{3} \end{aligned}$$

Termination

- 1. The *effective code rate*, $R_{\text{effective}}$, is defined as the average number of input bits carried by an output bit.
- 2. In practice, the input sequences are with finite length.
- 3. In order to terminate a convolutional code, some bits are appended onto the information sequence such that the shift registers return to the zero.
- 4. Each of the k input sequences of length L bits is padded with m zeros, and these k input sequences jointly induce n(L+m) output bits.
- 5. The effective rate of the terminated convolutional code is now

$$R_{\text{effective}} = \frac{kL}{n(L+m)} = R \frac{L}{L+m},$$

where $\frac{L}{L+m}$ is called the *fractional rate loss*.

- 6. When L is large, $R_{\text{effective}} \approx R$.
- 7. All examples presented are terminated convolutional codes.

Truncation

- 1. The second option to terminate a convolutional code is to stop for t > L no matter what contents of shift registers have.
- 2. The effective code rate is still R.
- 3. The generator matrix is clipped after the Lth column:



where $\boldsymbol{G}_{[L]}^c$ is an $(k \cdot L) \times (n \cdot L)$ matrix.

4. The drawback of truncation method is that the last few blocks of information sequences are less protected.



$$\begin{aligned} \boldsymbol{v} &= \boldsymbol{u} \cdot \boldsymbol{G}_{[5]}^c \\ &= (1, 1, 1, 0, 1) \cdot \begin{pmatrix} 11 & 10 & 11 & & \\ & 11 & 10 & 11 & \\ & & 11 & 10 & 11 \\ & & & 11 & 10 \\ & & & & 11 \end{pmatrix} \\ &= (11, 01, 10, 01, 00) \end{aligned}$$



$$oldsymbol{v} = oldsymbol{u} \cdot oldsymbol{G}_{[2]}^c$$
 $= (11,01) \cdot egin{pmatrix} 100 & 001 \ 010 & 001 \ & 100 \ & 010 \ \end{pmatrix}$
 $= (110,010)$

Tail Biting

- 1. The third possible method to generate finite code sequences is called tail biting.
- 2. Tail biting is to start the convolutional encoder in the same contents of all shift registers (state) where it will stop after the input of L information blocks.
- 3. Equal protection of all information bits of the entire information sequences is possible.
- 4. The effective rate of the code is still R.
- 5. The generator matrix has to be clipped after the Lth column

and manipulated as follows:





$$v = u \cdot \tilde{G}_{[5]}^{c}$$

$$= (1, 1, 1, 0, 1) \cdot \begin{pmatrix} 11 & 10 & 11 & \\ & 11 & 10 & 11 \\ & & 11 & 10 & 11 \\ 11 & & 11 & 10 \\ 10 & 11 & & 11 \end{pmatrix}$$

$$= (01, 10, 10, 01, 00)$$



$$\begin{aligned} \boldsymbol{v} &= \boldsymbol{u} \cdot \tilde{\boldsymbol{G}}_{[2]}^{c} \\ &= (11, 01) \cdot \begin{pmatrix} 100 & 001 \\ 010 & 001 \\ 001 & 100 \\ 001 & 010 \end{pmatrix} \\ &= (111, 010) \end{aligned}$$

References

[1] M. Bossert, *Channel Coding for Telecommunications*, New York, NY: John Wiley and Sons, 1999.